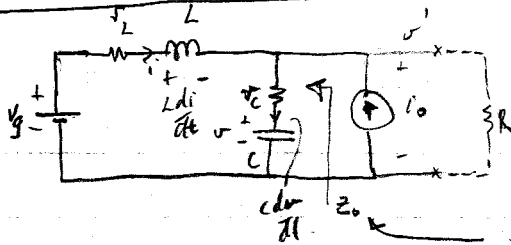


VOLTAGE MODE PROOF CONTROL ANALYSIS

DTs



THE ANALYSIS PRESENTED HERE REFERS TO THE PAPER "DESIGN CONSIDERATIONS FOR URM TRANSIENT RESPONSE BASED ON THE OUTPUT IMPEDANCE" by K. YAO et al. (1)

represents output impedance without R connected

$$L \frac{di}{dt} = -(r_c + r_L) i - v + v_g = r_c i_o$$

$$C \frac{dv}{dt} = i + i_o$$

$$v' = r_c i + v + r_c i_o$$

$$\begin{bmatrix} \frac{di}{dt} \\ \frac{dv}{dt} \end{bmatrix} = \underbrace{\begin{bmatrix} -\frac{(r_c + r_L)}{L} & -\frac{1}{L} \\ \frac{1}{C} & 0 \end{bmatrix}}_{A_1} \begin{bmatrix} i \\ v \end{bmatrix} + \underbrace{\begin{bmatrix} \frac{1}{L} & -\frac{r_c}{L} \\ 0 & \frac{1}{C} \end{bmatrix}}_{B_1} \begin{bmatrix} v_g \\ i_o \end{bmatrix}$$

$$A_1 = A_2 = A \quad B_2 = \begin{bmatrix} 0 & -\frac{r_c}{L} \\ 0 & \frac{1}{C} \end{bmatrix}$$

$$y = v' = \underbrace{\begin{bmatrix} r_c & 1 \end{bmatrix}}_{C_1} \begin{bmatrix} i \\ v \end{bmatrix} + \underbrace{\begin{bmatrix} 0 & r_c \end{bmatrix}}_{E_1} \begin{bmatrix} v_g \\ i_o \end{bmatrix}$$

$$C_1 = C_2 = C$$

$$E_1 = E_2 = E$$

$$z_o = \frac{v'}{i_o} = C (sI - A)^{-1} B_{i_o} + E_{i_o}$$

$$= \begin{bmatrix} r_c & 1 \end{bmatrix} \begin{bmatrix} s + \frac{r_c + r_L}{L} & \frac{1}{L} \\ -\frac{1}{C} & s \end{bmatrix}^{-1} \begin{bmatrix} -\frac{r_c}{L} \\ \frac{1}{C} \end{bmatrix} + r_c$$

$$\Delta = |sI - A| = s^2 + s \frac{(r_1 + r_2)}{L} + \frac{1}{LC} = \frac{1}{LC} (LCs^2 + s(r_1 + r_2)L + 1)$$

$$Z_o = \frac{1}{\Delta} \begin{bmatrix} r_2 & 1 \\ \frac{1}{C} & s + \frac{r_1 + r_2}{L} \end{bmatrix} \begin{bmatrix} -\frac{r_2}{L} \\ \frac{1}{L} \end{bmatrix} + r_2$$

$$= \frac{1}{\Delta} \begin{bmatrix} r_2 s + \frac{1}{L} & -\frac{r_2}{L} + s + \frac{r_1 + r_2}{L} \end{bmatrix} \begin{bmatrix} -\frac{r_2}{L} \\ \frac{1}{L} \end{bmatrix} + r_2$$

$$= \frac{1}{\Delta} \left[\left(\frac{r_2^2}{L} s - \frac{r_2}{LC} + \frac{s}{C} + \frac{r_2}{LC} \right) + \Delta r_2 \right]$$

$$= \frac{1}{\Delta} \left(-\frac{r_2^2}{L} s - \frac{r_2}{LC} + \frac{s}{C} + \frac{r_2}{LC} + r_2 LC s^2 + \frac{r_2^2}{L} s + \frac{r_2 r_2 s}{L} + \frac{r_2}{LC} \right)$$

$$= \frac{1}{\Delta} r_2 \left(\frac{r_2}{L} s^2 + \frac{s}{r_2 C} + \frac{r_2 s}{L} + \frac{1}{LC} \right)$$

$$= \frac{1}{\Delta} \frac{1}{LC} r_2 \left[\frac{r_2}{r_2} LC s^2 + \left(\frac{L}{r_2} + r_2 C \right) s + 1 \right]$$

$$= \frac{1}{LC} r_2 \left(1 + \frac{s}{\omega_L} \right) \left(1 + \frac{s}{\omega_C} \right)$$

$$= r_2 \frac{\left(1 + \frac{s}{\omega_L} \right) \left(1 + \frac{s}{\omega_C} \right)}{1 + \frac{s}{\omega_{w_0}} + \frac{s^2}{\omega_0^2}}$$

$$\omega_L = \frac{r_2}{L}$$

$$\omega_C = \frac{1}{r_2 C}$$

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

$$Q = \frac{\sqrt{LC}}{r_1 + r_2}$$

(3)

$$G_{rod}(s) = C (sI - A)^{-1} B_d + E_d$$

$$B_d = (A_1 - A_2) X + (B_1 - B_2) U = B_1 V_g \quad \text{since } A_1 = A_2$$

$$E_d = (C_1 - C_2) X + (E_1 - E_2) U = 0 \quad \text{since } C_1 = C_2 \text{ and } E_1 = E_2$$

$$= \begin{bmatrix} r_c & 1 \end{bmatrix} \begin{bmatrix} s + \frac{r_c + r_L}{L} & \frac{1}{L} \\ -\frac{1}{C} & s \end{bmatrix}^{-1} \begin{bmatrix} \frac{1}{L} \\ 0 \end{bmatrix} V_g \quad (I_o = 0)$$

$$\Delta = |sI - A| = \frac{1}{LC} (LCs^2 + s(r_c + r_L)c + 1)$$

$$G_{rod}(s) = \frac{1}{\Delta} \begin{bmatrix} r_c & 1 \end{bmatrix} \begin{bmatrix} s & -\frac{1}{L} \\ \frac{1}{C} & s + \frac{r_c + r_L}{L} \end{bmatrix} \begin{bmatrix} \frac{1}{L} \\ 0 \end{bmatrix} V_g$$

$$= \frac{1}{\Delta} \begin{bmatrix} r_c s + \frac{1}{L} & -\frac{r_c}{L} + s + \frac{r_c + r_L}{L} \\ \frac{1}{C} & s + \frac{r_c + r_L}{L} \end{bmatrix} \begin{bmatrix} \frac{1}{L} \\ 0 \end{bmatrix} V_g$$

$$= \frac{1}{\Delta} \left(\frac{r_c s}{L} + \frac{1}{LC} \right) V_g$$

$$= \frac{1}{\Delta} \frac{1}{LC} \left(1 + \frac{s}{\frac{1}{r_c C}} \right) V_g$$

$$G_{rod}(s) = V_g \frac{\left(1 + \frac{s}{\omega_c} \right)}{1 + \frac{s}{\omega_n} + \frac{s^2}{\omega_n^2}}$$

$$\omega_c = \frac{1}{r_c C}$$

$$\omega_n = \frac{1}{\sqrt{LC}}$$

$$Q = \frac{\sqrt{LC}}{r_c + r_L}$$

$$z_{oc} = \frac{z_o}{1+T} = \frac{z_L}{1+F_m G_{ud} G_{com}} = r_c$$

$$\Rightarrow r_c = \frac{r_L \left(1 + \frac{s}{\omega_L}\right) \left(1 + \frac{s}{\omega_C}\right)}{1 + \frac{s}{a\omega_o} + \frac{s^2}{\omega_o^2}}$$

$$1 + \frac{F_m \left(1 + \frac{s}{\omega_C}\right) V_g G_{com}}{1 + \frac{s}{a\omega_o} + \frac{s^2}{\omega_o^2}}$$

$$r_c = \frac{r_L \left(1 + \frac{s}{\omega_L}\right) \left(1 + \frac{s}{\omega_C}\right)}{\left(1 + \frac{s}{a\omega_o} + \frac{s^2}{\omega_o^2}\right) + F_m \left(1 + \frac{s}{\omega_C}\right) V_g G_{com}}$$

$$r_c \left(1 + \frac{s}{a\omega_o} + \frac{s^2}{\omega_o^2}\right) + r_c F_m V_g \left(1 + \frac{s}{\omega_C}\right) G_{com} = r_L + \left(\frac{1}{\omega_L} + \frac{1}{\omega_C}\right) r_L s + \frac{r_L s^2}{\omega_L \omega_C}$$

$$\begin{aligned} \frac{r_c}{\omega_o} + \frac{r_c s}{a\omega_o} + \frac{r_c s^2}{\omega_o^2} &\Rightarrow r_c F_m V_g \left(1 + \frac{s}{\omega_C}\right) G_{com} = \underbrace{r_L - r_c + \left[\left(\frac{1}{\omega_L} + \frac{1}{\omega_C}\right) r_L - \frac{r_c}{a\omega_o}\right] s + \left(\frac{r_L}{\omega_L \omega_C} - \frac{r_c}{\omega_o^2}\right) s^2}_{\left(1 + \frac{s}{\omega_{zv}}\right) \left(1 + \frac{s}{\omega_{zv}'}\right)} \quad (*) \\ &= (r_L - r_c) \left(1 + \frac{s}{\omega_{zv}}\right) \left(1 + \frac{s}{\omega_{zv}'}\right) \end{aligned}$$

$$\text{where } \omega_{zv} = \frac{b - \sqrt{b^2 - 4ac}}{2a} \quad \omega_{zv}' = \frac{b + \sqrt{b^2 - 4ac}}{2a}$$

$$\begin{aligned} \text{where } a &= \frac{r_L}{\omega_L \omega_C} - \frac{r_c}{\omega_o^2} \\ b &= r_L \left(\frac{1}{\omega_L} + \frac{1}{\omega_C}\right) - \frac{r_c}{a\omega_o} \\ c &= r_L - r_c \end{aligned}$$

(5)

$$\Rightarrow G_{con} = \frac{(r_L - r_C) \left(1 + \frac{s}{\omega_{zv}}\right) \left(1 + \frac{s}{\omega'_{zv}}\right)}{r_C F_m V_g \left(1 + \frac{s}{\omega_c}\right)}$$

this higher zero is ignored

$$\Rightarrow G_{con} = K_v \frac{1 + \frac{s}{\omega_{zv}}}{1 + \frac{s}{\omega_{pv}}}$$

$$K_v = \frac{r_L - r_C}{r_C F_m V_g}$$

$$\omega_{pv} = \omega_c$$

ω_{zv} is given above.

The paper "Design Considerations for VRM Transient Response Based on the Output Impedance" leaves the analysis at the above point, namely identifying a zero at ω_{zv} (as given above) and ignoring ω'_{zv} . However, the analysis can be made more complete and accurate as follows on the next page.

MORE COMPLETE AND ACCURATE ANALYSIS

We begin by copying equation (*) from two pages ago:

$$r_c F_m V_g \left(1 + \frac{s}{\omega_c}\right) G_{con} = r_L - r_c + \left[\left(\frac{1}{\omega_L} + \frac{1}{\omega_c} \right) r_L - \frac{r_c}{\omega \omega_0} \right] s + \left(\frac{r_L}{\omega_L \omega_c} - \frac{r_c}{\omega_0^2} \right) s^2 \quad (*)$$

$$\text{where } \omega_L = \frac{r_L}{L} \quad \omega_0 = \frac{1}{\sqrt{LC}}$$

$$\omega_c = \frac{1}{r_c C} \quad R = \frac{\sqrt{4L}}{r_L + r_c}$$

substituting these relations into the coefficient term of s^2 of (*) we have:

$$\frac{r_L}{\omega_L \omega_c} - \frac{r_c}{\omega_0^2} = \frac{r_L}{\frac{r_L}{L} \frac{1}{r_c C}} - \frac{r_c}{\frac{1}{LC}}$$

$$= r_c (LC - LC)$$

$$= 0$$

So the quadratic in s is reduced to first order in s , and the right hand side of (*) becomes

$$= r_L - r_c + \left[\left(\frac{L}{r_L} + r_c C \right) r_L - \frac{r_c}{\frac{\sqrt{4L}}{r_L + r_c} \frac{1}{\sqrt{LC}}} \right] s$$

$$= r_L - r_c + \left[L + r_c r_L C - r_c \left(\frac{r_L + r_c}{\sqrt{4L}} \right) C \right] s$$

$$= r_L - r_c + \left(L - r_c^2 C \right) s$$

$$= (r_L - r_c) \left(1 + \frac{L - r_c^2 C}{r_L - r_c} s \right)$$

7

$$= (r_h - r_c) \left(1 + \frac{s}{\omega_2} \right)$$

$$\text{where } \omega_2 = \frac{r_h - r_c}{L - r_c^2 C}$$

Together with the left hand side of (*) we have

$$r_c F_m V_g \left(1 + \frac{s}{\omega_c} \right) G_{con} = (r_h - r_c) \left(1 + \frac{s}{\omega_2} \right)$$

$$\Rightarrow G_{con} = \frac{(r_h - r_c) \left(1 + \frac{s}{\omega_2} \right)}{r_c F_m V_g \left(1 + \frac{s}{\omega_c} \right)}$$

$$= K_D \frac{\left(1 + \frac{s}{\omega_{2D}} \right)}{1 + \frac{s}{\omega_{pD}}}$$

$$\text{where } K_D = \frac{r_h - r_c}{r_c F_m V_g}$$

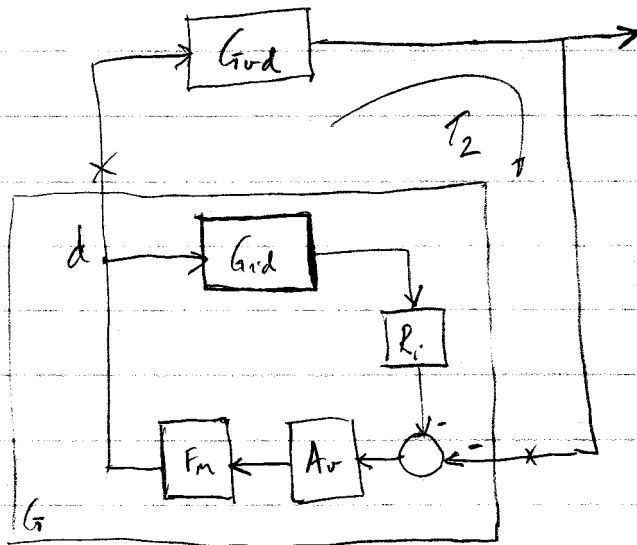
$$\omega_{pD} = \omega_c$$

$$\omega_{2D} = \frac{r_h - r_c}{L - r_c^2 C}$$

DERIVATION OF EQNS. (17) AND (18) IN THE PAPER

"OPTIMAL DESIGN OF THE ACTIVE DROOP CONTROL METHOD
FOR THE TRANSIENT RESPONSE"

(1)

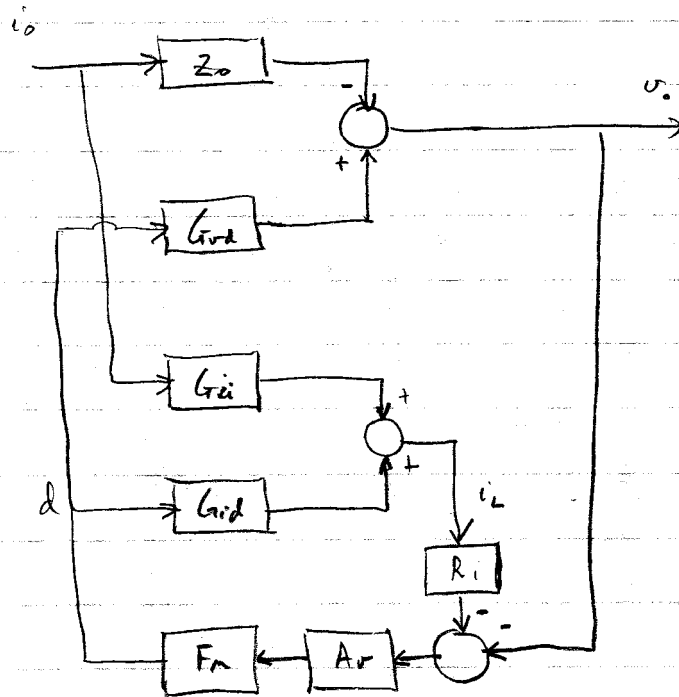


$$G = \frac{-F_m A_v}{1 + F_m A_v G_d R_i}$$

$$T_2 = G G_d$$

$$= \frac{-F_m A_v G_d}{1 + F_m A_v G_d R_i}$$

$$T_2 = \frac{-T_o}{1 + T_i} \quad (17)$$



$$v_o = -Z_o i_o + G_{ou} d$$

$$d = -F_m A_v (v_o + R_i i_L)$$

$$i_L = G_{ii} v_o + G_{id} d$$

$$d = -F_m A_v [v_o + R_i (G_{ii} v_o + G_{id} d)]$$

$$d = -F_m A_v v_o - F_m A_v R_i G_{ii} v_o - F_m A_v R_i G_{id} d$$

$$d (1 + F_m A_v R_i G_{id}) = -F_m A_v v_o - F_m A_v R_i G_{ii} v_o$$

$$v_0 = -Z_0 i_0 + G_0 d$$

$$d = \frac{-F_n A_0 v_0}{1 + F_n A_0 R_i G_{i0}} - \frac{F_n A_0 R_i G_{i0} i_0}{1 + F_n A_0 R_i G_{i0}}$$

$$= \frac{-F_n A_0 v_0}{1 + T_i} - \frac{F_n A_0 R_i G_{i0} i_0}{1 + T_i}$$

$$\Rightarrow v_0 = -Z_0 i_0 - \frac{F_n A_0 G_0 d v_0}{1 + T_i} - \frac{F_n A_0 G_0 R_i G_{i0} i_0}{1 + T_i}$$

$$v_0 = -\left(Z_0 + \frac{F_n A_0 G_0 R_i G_{i0}}{1 + T_i} \right) i_0 - \frac{T_0}{1 + T_i} v_0$$

$$\left(\frac{1 + T_0}{1 + T_i} \right) v_0 = \frac{-\left(Z_0 (1 + T_i) + F_n A_0 G_0 R_i G_{i0} \right) i_0}{1 + T_i}$$

$$Z_{oc} = \frac{v_0}{i_0} = \frac{-\left(Z_0 (1 + T_i) + F_n A_0 G_0 R_i G_{i0} \right)}{1 + T_i}$$

$$\frac{1 + T_0 + T_i}{1 + T_i}$$

$$= \frac{-\left(Z_0 (1 + T_i) + F_n A_0 G_0 R_i G_{i0} \right)}{1 + T_0 + T_i}$$

$$Z_{oc} = \frac{-\left(Z_0 (1 + T_i) + \overbrace{F_n A_0 G_0 R_i}^{T_i} \cdot \frac{G_0 G_{i0}}{G_0} \right)}{1 + T_0 + T_i}$$

$$1 + T_0 + T_i$$

EFFECT OF PARASITICS ON THE CONTROL-TO-OUTPUT
TRANSFER FUNCTION OF THE BUCK CONVERTER

$$\frac{\hat{v}}{\hat{d}}(s) = \frac{V_g}{1 + \frac{sL}{R} + s^2LC}$$

Effect of ESR of L : $\frac{sL}{R} \rightarrow \frac{sL}{sL+r_L}$

$$\frac{\hat{v}}{\hat{d}}(s) = \frac{V_g}{1 + \frac{sL+r_L}{R} + (sL+r_L)sC}$$

$$= \frac{V_g}{1 + \frac{r_L}{R} + \frac{sL}{R} + s r_L C + s^2LC}$$

$$\approx \frac{V_g}{1 + s\left(\frac{L}{R} + r_L C\right) + s^2LC} \quad \text{for } r_L \ll R$$

Effect of ESR of C : $\frac{1}{sC} \rightarrow \frac{1}{sC+r_C} = \frac{1+s r_C C}{sC}$

$$\frac{\hat{v}}{\hat{d}}(s) = \frac{V_g}{1 + \frac{sL}{R} + r_L \left(\frac{sC}{1+s r_C C}\right) + sL \frac{sC}{1+s r_C C}}$$

$$= V_g \frac{1 + s r_C C}{1 + s r_C C + \frac{sL}{R} (1 + s r_C C) + s r_L C + s^2LC}$$

$$= V_g \frac{1 + s r_C C}{1 + s \left[(r_C + r_L)C + \frac{L}{R} \right] + s^2LC \left(1 + \frac{r_C}{R}\right)}$$

$$\approx V_g \frac{1 + s r_C C}{1 + s \left[(r_C + r_L)C + \frac{L}{R} \right] + s^2LC} \quad r_C \ll R$$